MOTION OF A CONDUCTIVE PISTON IN A CHANNEL WITH VARIABLE INDUCTANCE

V. V. Polyudov, V. M. Titov, and G. A. Shvetsov

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The motion of a conductive piston in the channel of a magnetohydrodynamic (MHD) generator of the conduction type with compound electrodes is considered. Formulas are obtained for calculation of the energy characteristics of the pulse MHD generator for various operational regimes. It is shown that in an MHD generator at magnetic Reynolds number values $\text{Re}_{\text{m}} = \mu_0 \sigma u_0 l \gg 1$ (where μ_0 is the permeability of a vacuum, σ is the electrical conductivity of the piston, u_0 is the initial velocity, and l is the characteristic dimension), the energy transferred to an ohmic load may significantly exceed the values obtained in [1, 2]. Conditions for high-efficiency transformation of piston kinetic energy to electrical energy are considered for limiting values of the ratio of the latter to initial magnetic field energy in the generator channel.

In [1, 2] an analysis was performed of the energy characteristics of impulse MHD generators with $\mathrm{Re}_\mathrm{m} \gg 1$.

In [1] it was maintained that the change in magnetic field in the MHD generator channel with motion of a plasma with $\mathrm{Re}_m \gg 1$ leads to a maximum value of the braking pressure $2\mathrm{B}^2 / \mu_0$ (B₀ is the initial magnetic field).

It follows from this that the power and energy are then limited respectively to $4W_0u_0/l_0$ and $4W_0$ (W_0 is the initial magnetic field energy and l_0 is the generator length).

In [2], in analysis of a magnetocumulative generator with ohmic load it is concluded that the energy supplied to an active load R_1 = const has a maximum equal to $W_0 lnN$, where $N = L_0/L_1$ is the ratio of the initial generator channel inductance to the load inductance.

The conclusions of [1, 2] require refinements, as cited, for example, in [3-6], and there is no detailed analysis of the energy characteristics of impulse MHD generators in the literature. This present study will consider the basic energy relationships in impulse MHD systems using a simple electrotechnical model.*

1. We shall consider the operation of a linear MHD generator, with electrodes connected to an inductive-resistive load. These electrodes are switched to the primary energy source, a capacitor bank, and at the moment of maximum discharge current I_0 a moving conductive piston shorts the electrodes; the primary energy source is switched out of the circuit and will not be considered further. The inductance of the circuit at the switching time is L_0 , and for an arbitrary moment of time

$$L(t) = \begin{cases} L_0 - \int_0^t \dot{L}dt & (t \in [0; T]) \\ L_1 + L_2 & (t \geqslant T) \end{cases}$$

*V.S. Sokolov has called attention to the errors in the conclusion of [1].

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$$\int_{0}^{T} u(t) dt = l_{0}$$

L is the time rate of change of inductance, L_2 is the internal inductance of the generator, and L_1 is the load inductance.

The sum of the voltages and the energy balance for such a circuit have the form

$$\frac{d}{dt}(L(t)I(t)) + I(t)R = 0 \qquad (R = R_1 + R_2)$$
 (1.1)

$$\frac{L_0 I^{-2}}{2} + (E(0) - E(t)) = \frac{LI^2}{2} + \int_0^t I^2 R \, dt = W + \varepsilon$$
 (1.2)

where E is the kinetic energy of the moving piston, R_2 is the internal resistance of the generator, and R_1 is the load resistance.

In these expressions R_1 , R_2 , \dot{L} vary with time and are functions of the energy generated and the piston position within the generator. Therefore attaining solutions of Eqs. (1.1), (1.2) for the general case in the form of an exact expression in elementary functions is impossible. However, if it is assumed that the quantities mentioned are constant, Eq. (1.1) admits a simple solution and W and ϵ may be calculated.

Assuming

$$\dot{L} = \text{const}, R_1 = \text{const}, R_2 = \text{const}$$
 (1.3)

we find for $t \in [0; T]$

$$I = I_0 N(t)^{1-1/\gamma} \tag{1.4}$$

where N(t) = $L_0/L(t)$ is the coefficient of circuit inductance change, and $\gamma = \dot{L}/R_{\bullet}$

For t > T the circuit does not change ($\dot{L} = 0$) and the current is found from the equation

$$L_* dI/dt + IR = 0$$

$$I = I_* \exp\left[-RL^{-1}_*(t-T)\right] = I_0 N_*^{1-1/\gamma} \exp\left[-RL^{-1}_*(t-T)\right]. \tag{1.5}$$

The values of L, I, N, and W at time T are denoted by L_* , I_* , N_* , W_* .

It follows from Eq. (1.4) that the current in the circuit increases for $\gamma > 1$ and decreases for $\gamma < 1$, while the rate of decrease increases for $\gamma \to 0$.

For $t \in [0, T]$ let $\gamma > 1$, in which case three segments may be delineated in the curve of current with respect to time. The first phase up to t = 0 is determined by the primary energy source, which produces the current I_0 .

In the second phase from 0 to T the current in the circuit increases, and in the third phase for t > T the current decays exponentially.

The energy supplied to the load R1 after the piston shorts the generator electrodes will be

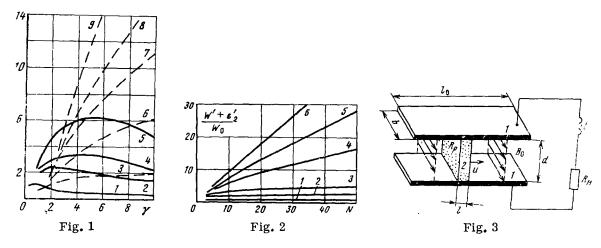
$$\epsilon' = \epsilon_2' - \epsilon_3'$$

The indices 2, 3 denote the respective phases of the process

$$\varepsilon_{2}' := \int_{\delta}^{\tau} I_{2}^{2} R_{1} dt := \begin{cases} W_{0} \frac{2k}{\gamma - 2} \left(N_{\bullet}^{1-2\gamma} - 1 \right) & (\gamma \neq 2) \\ k W_{0} \ln N_{\bullet} & (\gamma = 2) \end{cases}$$
 (1.6)

$$\varepsilon_{3}' = \int_{T}^{T \pm t_{3}} I_{3}^{2} R_{1} dt = kW_{\bullet}' \left[1 - \exp\left(-\frac{2R}{L_{\bullet}} t_{3} \right) \right] \quad (k = R_{1}/(R_{1} \pm R_{2}), W_{0} = L_{0} I_{0}^{2}/2)$$
(1.7)

$$W_{*}' = W_0 N^{1-2 \, Y} \tag{1.8}$$



where W*' is the magnetic field energy at t = T, and k is the load coefficient.

It is interesting that for arbitrary variation of $\dot{L}'(t)$ and R(t), the values of I_* , ϵ_2' and W_*' are still determined by Eqs. (1.4), (1.6), (1.8) if

$$L(t)/R(t) = \gamma = \text{const } t \in [0; T]$$

Figure 1 shows curves illustrating the dependence of ϵ_2'/W_0 (solid line) and W_*'/W_0 (dashes) on γ for given N_* and k=1, while Fig. 2 shows curves of $(W_*'+\epsilon_2')/W_0$ versus N_* for given γ and k=1.

In Fig. 1 each pair of curves (1, 3), (2, 6), (4, 7), and (5,8) corresponds to $N_* = 2$, 10, 20, 50.

In Fig. 2, curves 1, 2, 3, 4, 5, 6 correspond to γ equal to 0.1, 1, 2, 5, 10, 50.

From Eqs. (1.6)-(1.8) and Figs. 1, 2 it follows that for $\gamma \le 1$ the system energy does not exceed $2W_0$. For $\gamma \le 2$ $\varepsilon_2' \le W_0 \ln N_*$, $W_*' \le W_0$.

For a given N_* there exists a γ at which ϵ_2 ' is maximum, while for $\gamma > 2$, beginning at some N_* this maximum may significantly exceed W_0 ln N_* . For $\gamma \gg 2$ the system energy increases basically due to magnetic field energy. The increase in energy is then limited by the ratio of the initial inductance to the inductance L_* . We will now assume $L_1 \gg L_2$. For $R_1 \gg R_2$ an energy W_* ' may be transferred to the load if the energy extraction time in the third phase is greater than the relaxation time of the circuit L_1 , R_1 :

$$t_3 \geqslant L_1/R_1 = \tau \tag{1.9}$$

In this case $\epsilon_3' \approx W_*'$ and the total energy transferred to the load $R_1 = \text{const}$ is $\epsilon' \approx \epsilon_2' + W_*$.

2. We will consider the motion of a conductive piston in an MHD channel of constant cross section, located in an external magnetic field B_0 (1-electrodes, 2-conductive piston in Fig. 3). As before, we assume the conditions of Eq. (1.3) to be fulfilled. The sum of the voltages for this case will be

$$\frac{d}{dt}(LI) + IR = u_0 B_0 d \tag{2.1}$$

With consideration of the initial conditions (I $|_{t=0} = 0$) for $t \in [0; T]$ we obtain

$$I = \begin{cases} \frac{B_0 b \mu_0^{-1} \ln N(t)}{\frac{B_0 b}{\mu_0} \frac{1}{1 - 1/\gamma} \left(N_{(t)}^{1 - 1/\gamma} - 1 \right) & (\gamma = 1) \\ \frac{B_0 b}{\mu_0} \frac{1}{1 - 1/\gamma} \left(N_{(t)}^{1 - 1/\gamma} - 1 \right) & (\gamma = 1) \end{cases}$$
(2.2)

For t > T

$$I = I_* \exp(-RL_{\bullet}^{-1}(t-T))$$
 (2.3)

The energy transferred to the active load R₁ is

$$\boldsymbol{\varepsilon''} = \int\limits_0^{\mathbf{T}} I_2^2 R_1 \, dt + \int\limits_{\frac{T}{T}}^{\mathbf{T} + \mathbf{t_0}} I_3^2 R_1 \, dt = \varepsilon_2'' + \varepsilon_3''$$

(indices 2, 3 are used for uniformity of notation)

$$v_{2}^{"} = \begin{cases} 4W_{0}k \left(1 - \frac{\ln^{2}N_{\bullet} + 2\ln N_{\bullet} + 2}{2N_{\bullet}}\right) & (\gamma = 1) \\ 4W_{0}k \left[1 - 4\left(1 - \frac{1}{N_{\bullet}^{1/2}}\right) - \frac{1}{N_{\bullet}} + \ln N_{\bullet}\right] & (\gamma = 2) \\ \frac{2W_{0}k\gamma}{(\gamma - 1)^{2}} \left[\frac{\gamma}{\gamma - 2}(N_{\bullet}^{1 - 2/\gamma} - 1) - 2\gamma\left(1 - \frac{1}{N_{\bullet}^{1/\gamma}}\right) + \left(1 - \frac{1}{N_{\bullet}}\right)\right]_{\gamma \neq 2}^{\gamma \neq 1} \end{cases}$$

$$(2.4)$$

$$\varepsilon_3'' = kW_*'' \left[1 - \exp\left(-2RL_*^{-1}t_3\right)\right]$$
 (2.5)

$$W_*'' = \begin{cases} \frac{W_0 N_*^{-1} \ln^2 N}{(\gamma - 1)^2} & (\gamma - 1) \\ \frac{W_0 \gamma^2}{(\gamma - 1)^2} [N_*^{1 - 2/\gamma} - 2N_*^{-1/\gamma} + 1/N_*] & (\gamma \neq 1) \end{cases}$$
 (2.6)

where $W_0 = B_0^2 V_0 / 2 \mu_0$ is the initial magnetic field energy, and V_0 is the useful volume of the MHD channel.

Figure 4 shows the dependence of ϵ_2 "/ W_0 (solid line) and W_* "/ W_0 (dashes) on γ for given N_* and k=1. For ϵ_2 ", just as for ϵ_2 ', a maximum is characteristic, although attained at lower γ values. In absolute value ϵ_{2max} insignificantly exceeds ϵ_{2max} .

If t_3 satisfies the condition of Eq. (1.9) and $R_1\gg R_2$, then ϵ_3 " $\approx W_*$ " and the total energy supplied to the active load R_1 = const is equal to ϵ " = ϵ_2 " + W_* ".

Thus, in the methods considered to obtain a maximum ratio ε/W_0 it is first necessary to achieve transformation of the kinetic piston energy into electromagnetic energy at the maximum possible γ , and then at minimum possible γ (with limit $\gamma = 0$) to transfer this energy to the load. Since

$$\gamma = L/R = (1 - K) \mu_0 \sigma u_0 \delta = (1 - k) \operatorname{Re}_m \delta/l$$

all the values defined in Eqs. (1.4)-(1.8), (2.2)-(2.6) can be considered as functions of the magnetic Reynolds number $\text{Re}_{\mathbf{m}}$, calculated from the piston length l (δ is the effective depth of the current layer in the piston, $\delta \leq l$).

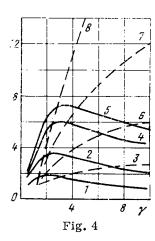
The analysis presented shows that in an impulse MHD generator with $\gamma \gg 1$ (Re_m = $\gamma(1-k)^{-1}$ $l/\delta \gg 1$) the energy transferred to an active load may significantly exceed the values obtained in [1, 2].

For $\mathrm{Re}_{\mathrm{m}} \ll 1$ the formulas for current (2.2) and energy (2.4) transform to the well-known expressions

$$I\leqslant B_0b\mu_0^{-1}\ (1-k)\ \mathrm{Re}_m$$

$$\varepsilon=2W_0k\left(1-k\right)\mathrm{Re}_m=Pl_0/u_0\qquad (P+\varepsilon zu_s^2B_0^2k\left(1-k\right)bdl)$$

where P is the power in the MHD generator load.



3. In real generators, as was noted above, $\dot{L} \neq const$, and so achieving solutions of Eqs. (1.1), (2.1) in complete form is not possible. However, for calculation, the energy conversion process can be represented as the sum of time intervals, during each of which $\dot{L} = const.$ In each interval the formulas for current and energy are analogous to those obtained above. It is evident that in systems in which \dot{L} is a decreasing function, for one and the same R_1 , R_2 , and N_*

$$\frac{\varepsilon \left[\gamma \left(0\right)\right]}{W_{0}} > \frac{\varepsilon \left[\gamma \left(t\right)\right]}{W_{0}} > \frac{\varepsilon \left[\gamma \left(T\right)\right]}{W_{0}}$$

In the general case \dot{L} is a function of piston speed and the geometrical dimensions of the generator channel. For a generator of constant cross section, the condition \dot{L} = const is equivalent to the condition that $u(t) \approx u_0$ = const.

It is evident that if the piston speed during the energy conversion process is almost constant, only a relatively small fraction of the kinetic energy will be transformed into electrical, and so the useful-work coefficient of the generator

$$\eta = (E_0 - E_*)/E_0$$

will be low. With decrease in piston speed η increases, but this leads to decrease in \dot{L} and finally to decrease in ϵ/W_0 . The problem of simultaneous maximization of η and ϵ/W_0 will not be considered here. We simply note that this requirement may be achieved if the reduction in piston speed is compensated by a corresponding change in the geometrical dimensions of the generator channel.

For transfer of energy W_* to an active load the condition of Eq. (1.9) is necessary. It may be satisfied in several manners. Let us assume that the piston shorts the circuit L_1R_1 , having reached the ends of the electrodes. If $l/u_* \ge \tau$, then $\epsilon_3 \approx W_*$. If the circuit L_1 , R_1 is closed for a period of time $t_3 < L_1'/R_1$, the energy W_* may be transferred to the load by artificially reducing the system relaxation time. For example, the load resistance may be increased sharply [2, 7]. The condition $t_3 \ge \tau R_1/R_3$ will then determine the necessary resistance increase.

The formulas presented in this study permit evaluation of the basic energy characteristics and analysis of the operation of impulse energy converters.

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